

same cable thickness and thus the similar fraction of cable insulation is used in each layer. The grading is achieved by changing the cable width by using a different number of wires (strands) in the cable. Since the wire diameter is not changed a greater flexibility in design is possible – the relative grading can be varied by varying the cable width (by using a different number of wires). The present design has 40 strands in the inner and 26 in the outer two layers.

TABLE I

THE MAJOR PARAMETERS OF THE PRESENT DESIGN FOR A 40 MM APERTURE COMMON COIL DESIGN MAGNET (IN SECOND COLUMN THE PARAMETERS FOR A 10 MM APERTURE CONFIGURATION ARE GIVEN IN PARENTHESIS).

Coil aperture (mm)	40 (10)
Number of layers	3
Computed quench field at 4.2 K (T)	13.8 (16.2)
Peak Fields, inner & outer layers (T)	15.0 & 10.5 (16.3 & 11.4)
Quench current, inner & outer layers (kA)	12.0 & 12.0 (8.7 & 13.0)
Wire Non-Cu J_{sc} {4.2 K, 12 T} (A/mm ²)	2000
Strand diameter (mm)	0.8
No. of strands, inner & outer layers	40, 26
Cable width, inner & outer layer (mm)	16.5, 10.7
Cu/Non-Cu ratio, inner & outer	0.7, 1.7
No. of turns (total)	120
Height of each layer (mm)	70
Bore spacing (mm)	170
Minimum coil bend radius (mm)	50
Yoke outer radius (mm)	250

The center-to-center spacing between the two apertures is 170 mm and height of each layer is 70 mm. This gives a minimum coil bend radius (in the ends) of 50 mm, which is about an order of magnitude more than that in conventional block designs or in cosine (θ) designs. The width of the iron insert between the two apertures is 80 mm. The iron insert reduces the coupling (cross-talk) between the two apertures and gives a higher short sample field. The iron insert, however, generates large saturation induced harmonics in this geometry. The issue of iron insert and the optimization of saturation induced harmonics will be examined later.

The yoke outer radius is only 250 mm. The compact nature of the design is related to the magnet geometry. When this design is adopted for a 50 mm aperture magnet and compared with D20 (a similar field and same aperture dipole built and tested earlier at LBNL [5]), the yoke or coldmass area per aperture in this design is only a quarter of that in D20.

III. MAGNETIC DESIGN OPTIMIZATION WITH ROXIE

The computer code ROXIE [6, 7] will be used in optimizing the field quality and magnetic design of an accelerator magnet built on this concept. This program has been used in designing a number of LHC and other magnets based on a

cosine (θ) coil geometry. ROXIE has now been upgraded to optimize a magnet built with racetrack coils inside an arbitrarily shaped iron aperture. ROXIE can also minimize the end harmonics in such a geometry.

The primary parameters in coil cross-section optimization are (a) internal spacers within each coil layer (similar to the wedges in cosine θ magnets) and (b) coil height and beginning position of each layer (similar to the pole angle). The coil geometry must have an up-down asymmetry with respect to the midplane in each aperture to compensate for the coupling between the two apertures. The shape and dimensions of various end spacers and length of coil are used as parameters to minimize both normal and skew end harmonics. The iron saturation is minimized by shaping the iron inner surface between the two apertures and around the coil modules. In addition, the holes and cutouts (in the iron cross section) and the yoke outer surface can be used as parameters in minimizing saturation induced normal and skew harmonics. A computer model of the present design (which is not primarily optimized for field quality) is shown in Fig. 2.

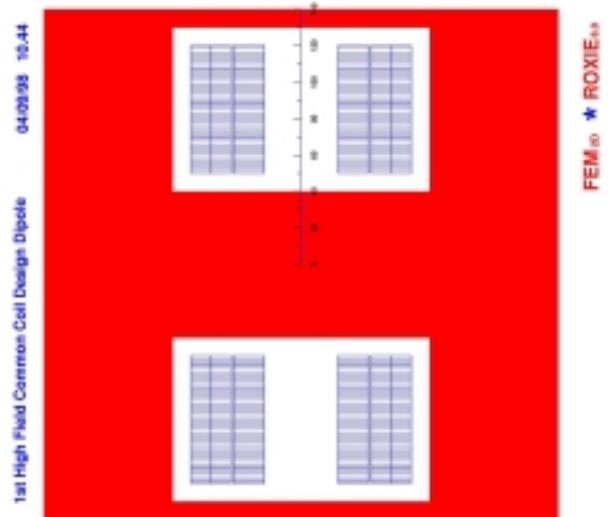


Fig. 2. A ROXIE model of the high field common coil design magnet. The figure shows only a part of the iron and 3-layer of coils with iron inserts between the two apertures.

IV. MECHANICAL DESIGN

At fields of 14 to 16 Tesla the mechanical design is driven by two requirements: supporting high Lorentz forces and managing coil stresses. The integrated Lorentz forces act to push the magnet coils apart, thereby requiring a rigid support structure design to minimize coil displacements and maintain magnet integrity. Within the coil winding the Lorentz forces act to compress the winding (transverse to winding direction); a stiff and strong coil design is required to minimize coil deformation and withstand internal stresses. In addition to structural concerns, thermal conductance is an issue since adequate heat transfer is required for thermal stability. Lastly, the mechanical design is based on criteria applicable to this R&D program: flexibility, modularity, short turnaround time, and ease of fabrication.

The current mechanical design is a “dedicated component” design. The magnet is separated into three main components and each serves a dedicated function. The magnet core resides within the iron yoke and serves to package the parts within the core (coils, islands, end shoes, spacers, etc.), the iron yoke surrounds the magnet core and serves as a flux return, and an external wire wrap is used around the iron yoke to provide preload and structural support. In contrast to the dedicated component design is an “integrated component” design wherein each component serves multiple functions (e.g. the function of structural support is shared between magnet core, iron yoke, and wire wrap). The dedicated component design provides flexibility and modularity appropriate for R&D magnets. Since the magnet core does not incorporate any support function, different magnet core designs can be used with the same external support (provided the support is adequately strong and rigid). Such a system lends itself to a program of systematic investigations. On the other hand, an integrated component design is inherently more efficient and is suited for production magnet designs.

A. External Support Design

Figures 3a and 3b show the horizontal and vertical force densities in a coil cross section for one quadrant of the magnet winding (with only main coils). At 14 T bore field, the total integrated horizontal force is 19 MN per meter of magnet straight section length (137 MPa distributed over the racetrack edge area); total integrated vertical force is 1.6 MN/m (21 MPa distributed over the racetrack face area). Total integrated forces are the main factor in designing the external magnet support which needs to be strong enough to support the integrated outward forces.

An external wrap of stainless steel wire has been successfully used for applying preload and support in a high field cosine (θ) magnet [5]. The current design uses a similar wire wrap technique for support. The wire is wound onto a bobbin placed around the iron yokes and locked into the bobbin after each wrap layer. Shaping the outer yoke perimeter provides control of horizontal and vertical preload. Design of the yoke gaps also determines effective preload and support. The wire wrap has two roles: vertical preload on the coils and support against the Lorentz forces.

In the vertical direction the coils are preloaded to a level which prevents the coil’s inner turn from separating from the island when the magnet is energized. Magnet training is expected to be reduced by preventing energy release due to coil movement. In the horizontal direction the preload is small. The preload ensures contact between the coil outer surface and support structure, but at high fields the inner coil surface is free. The accumulated Lorentz forces act mostly in the horizontal direction. The forces push on the iron yoke, which transfers the load to the external wire wrap support structure.

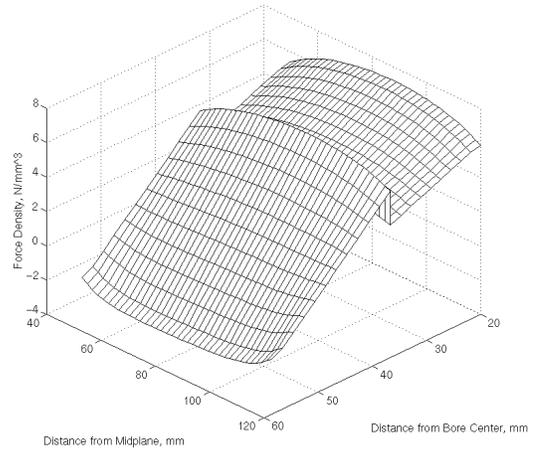


Fig. 3a. Horizontal force density

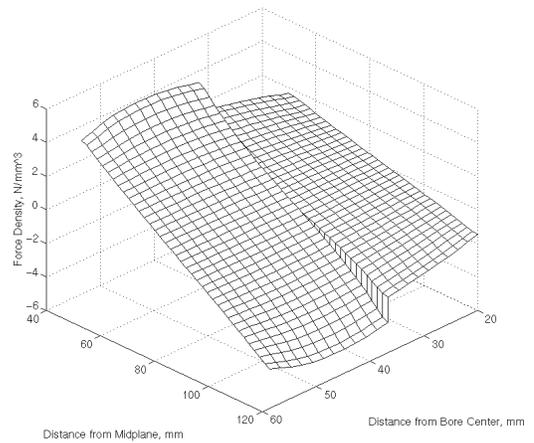


Fig. 3b. Vertical force density

Fig. 3. Horizontal and vertical force densities at 14 T bore field for coil cross section in quadrant 1 of high field common coil magnet. Positive values denote forces outward from magnet center.

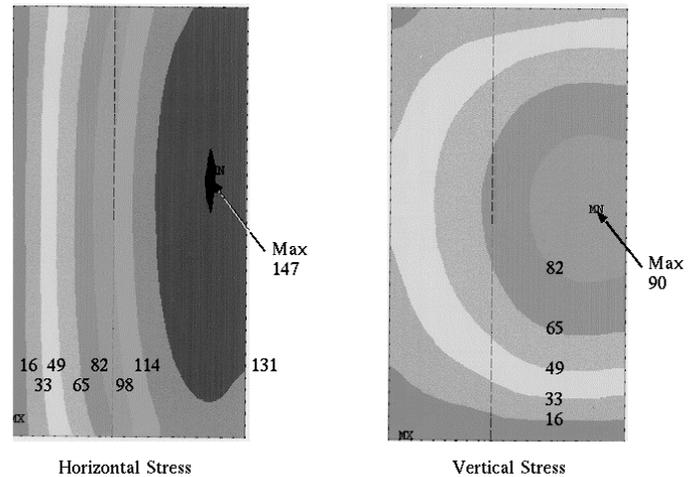


Fig. 4. Horizontal and vertical coil stresses for coil cross section in quadrant 1 of high field (14 T) common coil magnet. All values are compressive in MPa.

B. Coil Stresses

The force distribution *within* a coil cross section is the source of internal coil stresses when the magnet is energized. Note that for the common coil design the forces are low near the cross section center and increase (in opposite directions) towards the sides, effectively compressing the coil in both horizontal and vertical directions (see Fig. 4).

Epoxy impregnation is used to provide support for the brittle superconductors. The extra support provided by the epoxy allows normal handling of the heat treated brittle coils. The epoxy also serves to increase the mean stiffness of the composite coil. Increasing the stiffness results in lower deformation of the coil winding under Lorentz forces. The elastic modulus of epoxy impregnated Nb₃Sn cable in compression has been measured at up to 53 GPa transverse to the winding direction [8].

If no structural elements are incorporated into the coil winding, the stresses accumulate to very high levels (up to 150 MPa for this high field common coil design). A 16 Tesla dipole magnet is currently under development at Texas A&M University which uses a sophisticated rib and plate structure to provide structural support within the coil cross section [9]. The common coil design is amenable to similar stress management schemes and these may be incorporated later, when and if necessary. However, previous test results of an epoxy impregnated Nb₃Sn dipole indicate that the high coil stresses do not significantly degrade conductor performance up to 13.5 Tesla with comparable stress levels [5]. For the current common coil design, no structural support is incorporated into the coil winding. Moreover, it may be pointed out that although the stress accumulates to high levels in the coil, the high stresses occur at low field locations (see Figs. 1 and 4).

V. MAGNET R&D PROGRAM

The design approach described here allows a systematic and efficient magnet R&D program. One major design and performance issue in very high field Nb₃Sn magnets is the amount of cable degradation due to large stress accumulation. The *“in-magnet”* situation is better since the place of the highest stress accumulation is also the place of the highest field margin. Moreover, tests done on small cable samples to determine the degradation do not always simulate the actual magnet conditions. This design allows simulating *“actual magnet test situation”* by creating a field of 16.2 tesla (computed short sample assuming no stress degradation) in the 10 mm aperture mode (see Table 1) up from 13.8 tesla in the nominal 40 mm aperture mode. This is accomplished by changing/modifying some of the internal support modules inside the iron yoke and using two power supplies. A similar modification in internal structure allows addition and/or replacements of coils made with different size conductor.

A large bend radius in the magnet ends allows this design to accommodate coils made with High Temperature Superconductors (HTS). The advantage of HTS in very high field magnets is the small decrease in the critical current density with increasing field. The current density in HTS is not as yet

large enough to create a high field in a practical design but is sufficient to carry out a magnet R&D. An HTS insert coil in a hybrid magnet where the majority of the field comes from conventional low temperature superconductor (e.g. Nb₃Sn) will be a first important step towards the future HTS-based accelerator magnet research. This will address the relevant magnetic and mechanical design issues since here the HTS will be in an environment similar to that in a complete HTS magnet.

An overall modular approach also facilitates a systematic study of magnetic and mechanical design. In the past many such changes required building a new magnet which takes a long time. Moreover, a new magnet may not always reproduce a similar situation, since more than one parameter gets changed, inadvertently. In addition, the horizontal and vertical supports, which differ significantly in their design philosophies, can be studied separately. The same modules can also be used to carry out the high field magnet R&D for the proposed muon collider [10].

VI. CONCLUSIONS

A preliminary design of a high field 2-in-1 dipole based on a common coil design has been presented. The construction of this magnet follows the completion of the moderate field magnet now in preparation for test at LBNL. Given a relatively long lead time available for the next large hadron collider, the design philosophy and approach takes advantage of carrying out a systematic and innovative magnet R&D program to produce low cost magnets for an overall lower cost hadron colliders.

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